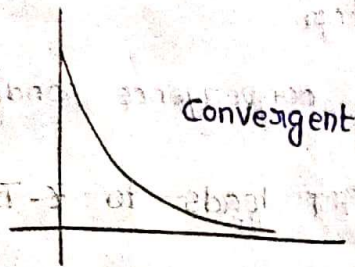
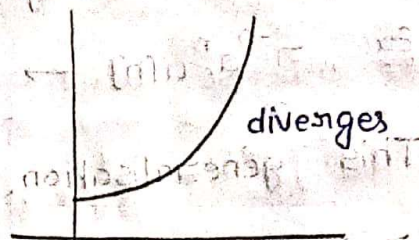


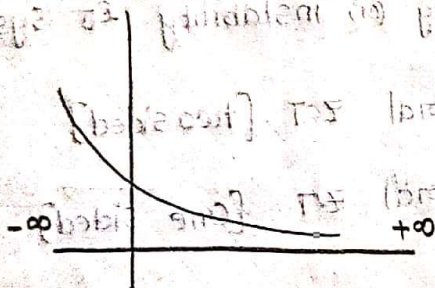
- Z-Transform in discrete time is analogous to Laplace transform in continuous time.
- F.T does not exist for not absolutely summable signals.



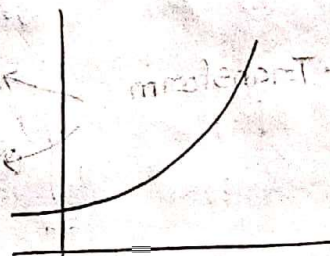
Finite, Abs summable



Not abs Summable



Not absolutely Summable

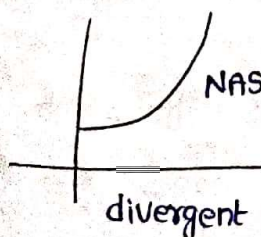


Not abs. Summable

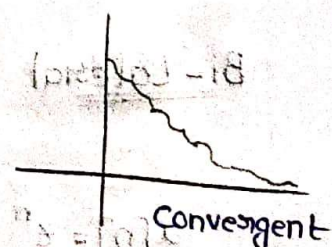
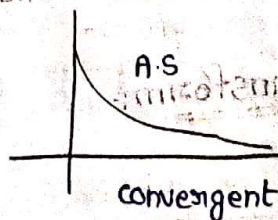
Absolutely Summable

 $\Rightarrow$ DTFT exists

Not Absolutely Summable

 $\Rightarrow$ DTFT doesn't exist

X



- Inputs: $a^n u[n]$; $a > 1$

$$a^{-n} ; -\infty < n < \infty$$

$u[n]$ and other signals which are not absolutely summable.

- Similarly, DTFT does not exist for signals that are not absolutely summable.

* Generalising the DTFT, so that the signal $x[n]$ is expressed as the sum of complex exponentials, z^n where $z = re^{j\omega}$

* This is equivalent to multiplying the signal by an exponential convergent factor.

Ex: $r^{-n} a^n u[n] \rightarrow$ Satisfies convergence condition for $r > |a|$

* This generalisation of DTFT leads to Z-Transform

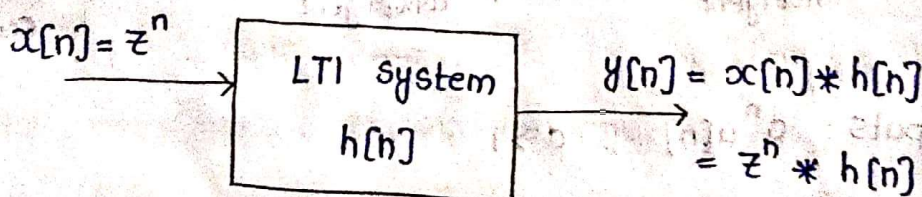
* Z-Transform $\rightarrow$ Analysis on unstable systems
[stability or instability of systems]

* Z-Transform $\rightarrow$ Bi-lateral Z.T [two sided]
 $\rightarrow$ Uni-lateral Z.T [One sided]

* Bi-lateral Z.T [nature of system characteristics such as stability, causality, freq response]

* Uni-lateral Z.T [Solving difference equations with initial conditions]

Bi-Lateral Z-Transform:



Consider applying a complex exponential input $x[n] = z^n$ to an LTI system with impulse response $h[n]$.

System o/p $y[n] = h[n] * x[n]$

$$y[n] = \sum_{k=-\infty}^{\infty} h[k] \cdot x[n-k]$$

$$= \sum_{k=-\infty}^{\infty} h[k] \cdot z^{n-k}$$

$$= z^n \sum_{k=-\infty}^{\infty} h[k] \cdot z^{-k}$$

$$= z^n \underbrace{\sum_{k=-\infty}^{\infty} h[k] \cdot z^{-k}}_{= H[z]}$$

$$y[n] = H[z] \cdot z^n$$

$$\text{where, } H[z] = \sum_{k=-\infty}^{\infty} h[k] z^{-k}$$

$$\text{Equivalently, } H[z] = \sum_{n=-\infty}^{\infty} h[n] z^{-n}$$

↓
Transfer function / System function of the LTI system.

Note: A signal for which the system output is a constant times the input is referred to as an eigenfunction of the system and the amplitude factor is referred to as the system Eigen value.

$z^n \Rightarrow$ Eigen function of LTI system

$H[z] \Rightarrow$ Eigen value.

$$\begin{aligned} * \text{Z.T. } H[z] &= \sum_{n=-\infty}^{\infty} h[n] z^{-n} \quad \text{Bi-lateral} \\ H[z] &= \sum_{n=0}^{\infty} h[n] z^{-n} \quad \text{Uni-lateral} \end{aligned} \quad \left. \vphantom{\sum_{n=-\infty}^{\infty}} \right\} \textcircled{1}$$

Inverse Z-Transform:-

Substituting $z = re^{j\omega}$ in $H[z]$ equation and using 'n' as the variable of summation.

$$H(z e^{j\omega}) = \sum_{n=-\infty}^{\infty} h[n] (z e^{j\omega})^{-n}$$

$$= \sum_{n=-\infty}^{\infty} [h[n] (z^{-n})] e^{-j\omega n}$$

$$H(z e^{j\omega}) = \text{DTFT} \{h[n] z^{-n}\}$$

$\therefore$ The inverse DTFT of $H(z e^{j\omega})$ must be $\{h[n] z^{-n}\}$

$$\therefore h[n] z^{-n} = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(z e^{j\omega}) e^{j\omega n} d\omega$$

$$h[n] = z^n \cdot \frac{1}{2\pi} \int_{-\pi}^{\pi} H(z e^{j\omega}) e^{j\omega n} d\omega$$

$$h[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(z e^{j\omega}) (z e^{j\omega})^n d\omega \rightarrow (2)$$

A change of variables is performed by letting

$$z = z e^{j\omega}$$

$$dz = j z e^{j\omega} d\omega$$

$$dz = j z d\omega$$

$$\Rightarrow d\omega = \frac{1}{j} z^{-1} dz$$

$\omega \rightarrow 0$ to 2π interval

which corresponds to one traversal around the circle $|z| = 1$

$$\therefore h[n] = \frac{1}{2\pi j} \oint_C H(z) z^{n-1} dz \quad \text{Inverse of Z.T.} \rightarrow (8)$$

'C' $\Rightarrow$ closed contour in the region of convergence (ROC) of $H(z)$ in anti-clockwise direction.

Note:-

For an arbitrary signal $x[n]$, the z-transform is defined as

$$z\{x[n]\} = X[z] = \sum_{n=-\infty}^{\infty} x[n] \cdot z^{-n} \rightarrow (4)$$

Inverse z.T of $X[z]$ is defined as

$$x[n] = \frac{1}{2\pi j} \oint X[z] z^{n-1} dz \rightarrow (5)$$

* The transform relationship between $x[n]$ & $X[z]$ as

$$x[n] \longleftrightarrow X[z]$$

Region of Convergence: The range till z-transform exists.

Relation b/w z.T & DTFT:

Discrete time signal $x[n]$

$$z\{x[n]\} = X[z] = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

Substituting $z = re^{j\omega}$ in the above equation,

$$\begin{aligned} z\{x[n]\} &= \sum_{n=-\infty}^{\infty} x[n] [re^{j\omega}]^{-n} \\ &= \sum_{n=-\infty}^{\infty} [x[n] r^{-n}] e^{-j\omega n} \end{aligned}$$

$$z\{x[n]\} = \mathcal{F}\{x[n] r^{-n}\}$$

$\therefore$ The z.T of $x[n]$ is the DTFT of $x[n] r^{-n}$

* If $r=1$

$$z\{x[n]\} = \mathcal{F}\{x[n]\} \text{ for } r=1.$$

The z.T reduces to the DTFT when $|z|=1$.

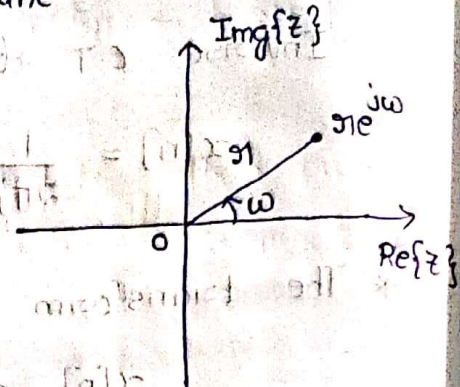
z-plane:

Complex number $z \Rightarrow$ Complex plane

$$z = re^{j\omega}$$

where, $r =$ distance from origin

$\omega =$ Angle from real axis

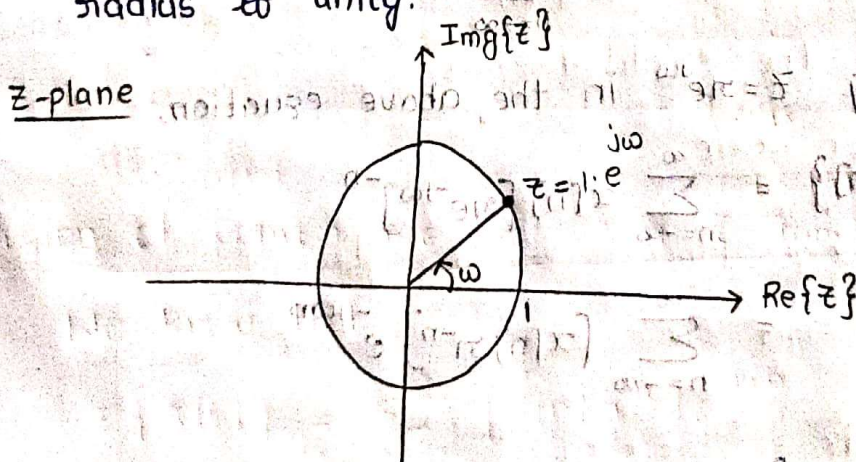


If $r=1 \Rightarrow$ z.T = DTFT

z-plane

$$X[e^{j\omega}] = X[z] \Big|_{z=e^{j\omega}}$$

Note:- The z.T reduces to the DTFT on the contour in the complex z-plane corresponding to a circle with a radius of unity.



The DTFT corresponds to the z-transform evaluated on the unit circle.

Poles and Zeros:-

General form of z-transform is the ratio of two polynomials in 'z'.

$$H[z] = \frac{b_m z^m + b_{m-1} z^{m-1} + \dots + b_1 z + b_0}{a_n z^n + a_{n-1} z^{n-1} + \dots + a_1 z + a_0}$$

In terms of factors of the polynomials in the numerator and denominator and to write the transfer function in terms of those factors.

$$H[z] = \frac{N[z]}{D[z]} = \frac{k(z-z_1)(z-z_2)(z-z_3)\dots(z-z_{m-1})(z-z_m)}{(z-p_1)(z-p_2)\dots(z-p_{n-1})(z-p_n)} \quad \text{--- (6)}$$

$$'o' \Leftarrow z \Rightarrow \text{Zeros}$$

$$'x' \Leftarrow p \Rightarrow \text{Poles}$$

Note:- Where the numerator & denominator polynomials $N[z]$ and $D[z]$ have real coefficients and

$$k = \frac{b_m}{a_n}$$

* z_i 's are the roots of the equation $N[z] = 0$ and are defined as "zeros".

* p_i 's are the roots of the equation $D[z] = 0$ and are defined as "poles".

In eqn (6), the factors in the numerator & denominator are written so that when $z = z_0$, the numerator $N[z] = 0$ and the transfer function vanishes.

$$\lim_{z \rightarrow z_0} H[z] = 0$$

Similarly when $z = P_i$ the denominator $D[z] = 0$ and the transfer function becomes infinity / unbounded.

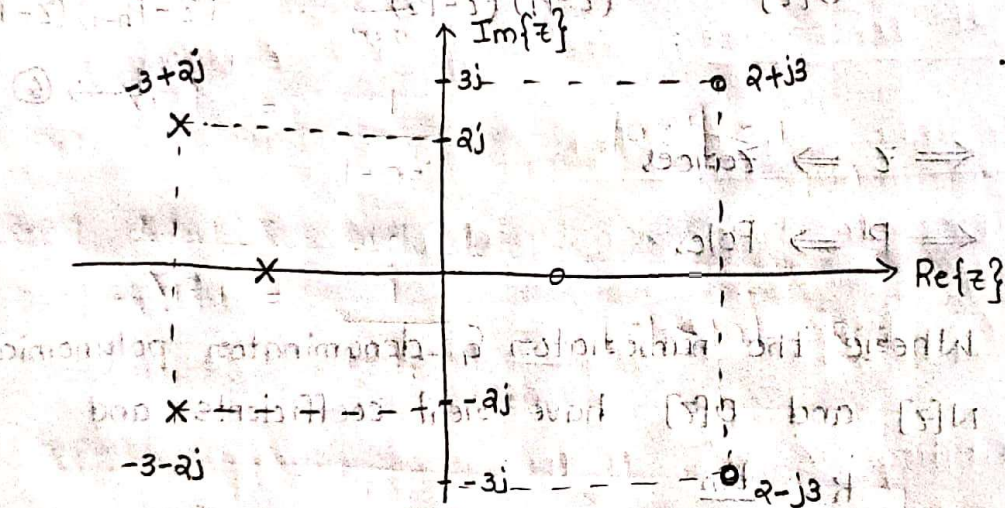
$$\lim_{z \rightarrow P_i} H[z] = \infty$$

Note:- All of the coefficients of polynomial $N[z]$ and $D[z]$ are real.

⇒ Poles and zeroes must be either purely real or appear in complex conjugate pairs.

⇒ Zeroes $z = 1, 2 \pm j3$

Poles $P = -2, -3 \pm j2$



Region of Convergence [ROC] for Z-Transform:

We know that,

$$z\{x[n]\} = \sum_{n=-\infty}^{\infty} [x[n]z^{-n}] e^{-j\omega n}$$

Z.T is guaranteed to converge

if $x[n]z^{-n}$ is absolutely summable.

i.e,
$$\sum_{n=-\infty}^{\infty} |x[n]z^{-n}| < \infty$$

This guarantees that $x[z]$ is finite,

$$Z\{x[n]\} = X[z] = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

substituting $z = re^{j\omega}$ in above equation,

$$X[z] = \sum_{n=-\infty}^{\infty} x[n] (re^{j\omega})^{-n}$$

$$X[z] = \sum_{n=-\infty}^{\infty} [x[n] r^{-n}] e^{-j\omega n}$$

$$|X[z]| = \left| \sum_{n=-\infty}^{\infty} [x[n] r^{-n}] e^{-j\omega n} \right|$$

$$|X[z]| \leq \sum_{n=-\infty}^{\infty} |x[n] r^{-n}| \underbrace{|e^{-j\omega n}|}_{=1}$$

$$|X[z]| \leq \sum_{n=-\infty}^{\infty} |x[n] r^{-n}|$$

So if $\sum_{n=-\infty}^{\infty} |x[n] r^{-n}| < \infty$ then $|X[z]| < \infty$

* The range of values of $|z| = r$ for which the Z.T converges is termed as "Roc".

Note:- Roc consists of those values of 'r' for which the DTFT of $x[n] r^{-n}$ converges.

⊛ Z-T exists for some signals that do not have a DTFT. But by limiting to certain range of values of 'r' we may ensure that $x[n] r^{-n}$ is absolutely summable, even though $x[n]$ is not absolutely summable by itself.

⊛ Roc provides the information about whether $x[n]$ is Fourier transformable or not.

∴ DTFT is obtained from bilateral Z-transform by setting " $r=1$ " the Roc in this case is a unit circle.

* If the ROC for $X(z)$ includes the unit circle, $x[n]$ is fourier transformable.

(Pb) $x[n] = a^n u[n]$

$$z \{x[n]\} = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

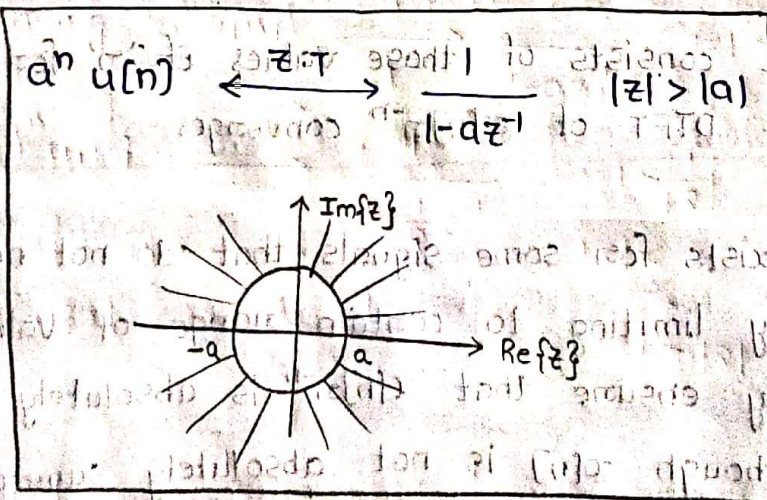
$$= \sum_{n=0}^{\infty} a^n u[n] z^{-n}$$

$$= \sum_{n=0}^{\infty} a^n z^{-n}$$

$$= \sum_{n=0}^{\infty} (az^{-1})^n$$

$$X(z) = \frac{1}{1-az^{-1}}, \quad |az^{-1}| < 1$$

$$X(z) = \frac{1}{1-az^{-1}}, \quad |z| > |a| : \text{ROC}$$



(Pb) $x[n] = -a^n u[-n-1]$

$$z \{x[n]\} = X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

$$u[n] = \begin{cases} 1 & n \geq 0 \\ 0 & n < 0 \end{cases}$$

$$u[-n-1] = \begin{cases} 1 & (-n-1) \geq 0 \Rightarrow n \leq -1 \\ 0 & (-n-1) < 0 \Rightarrow n > -1 \end{cases}$$

$$\begin{aligned} z \{x[n]\} = X[z] &= \sum_{n=-\infty}^{\infty} -a^n u[-n-1] \cdot z^{-n} \\ &= - \sum_{n=-\infty}^{-1} a^n z^{-n} \\ &= - \sum_{n=-\infty}^{-1} (az^{-1})^n \end{aligned}$$

$$X[z] = - \sum_{n=1}^{\infty} (a^{-1}z)^n = - \sum_{n=1}^{\infty} (a^{-1}z)^n$$

$$\sum_{n=1}^{\infty} \alpha^n = \frac{\alpha}{1-\alpha}, \quad |\alpha| < 1$$

$$\text{for } |a^{-1}z| < 1 \Rightarrow |z| < |a|$$

$$\frac{-z/a}{1-z/a}$$

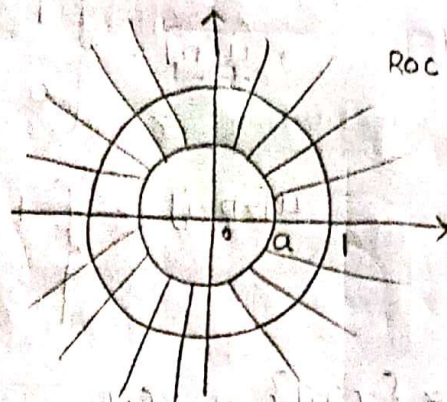
$$X[z] = \frac{-z}{a-z} ; |z| < |a| \Rightarrow X[z] = \frac{z}{z-a} ; |z| < |a|$$

$a^n u[n]$	$\longleftrightarrow$	$\frac{1}{1-az^{-1}}$	$ z > a $
$-a^n u[-n-1]$	$\longleftrightarrow$	$\frac{z}{z-a} = \frac{1}{1-az^{-1}}$	$ z < a $

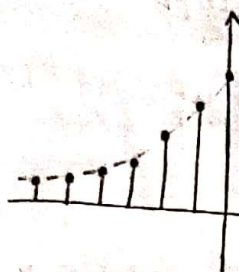
* $x[n] = a^n u[n] \quad 0 < a < 1$



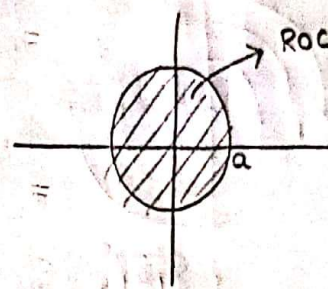
$\longleftrightarrow$



* $-a^n u[-n-1] \longleftrightarrow \frac{1}{1-az^{-1}} ; |z| < |a|$



$\longleftrightarrow$



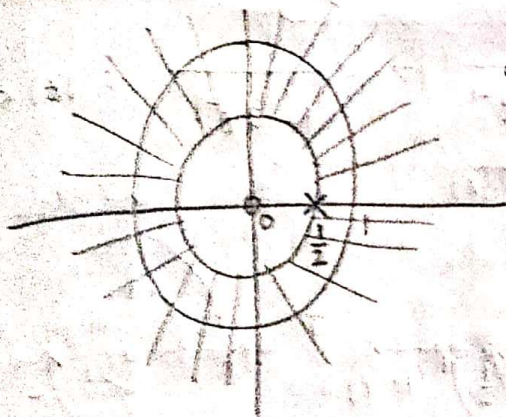
(Pb) Determine the z -Transform $x[n] = (\frac{1}{2})^n u[n] + 2^n u[n]$ and depict Roc and location of poles & zeroes in the z -plane.

Ans:- $z\{x[n]\} = z\{(\frac{1}{2})^n u[n]\} + z\{2^n u[n]\}$
By using $a^n u[n] \longleftrightarrow \frac{1}{1-az^{-1}} ; |z| > |a|$

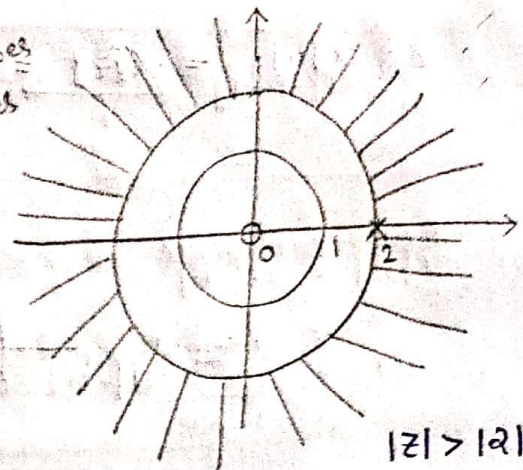
① $(\frac{1}{2})^n u[n] \longleftrightarrow \frac{1}{1-\frac{1}{2}z^{-1}} ; |z| > \frac{1}{2}$

② $2^n u[n] \longleftrightarrow \frac{1}{1-2z^{-1}} ; |z| > 2$

$X[z] = \frac{1}{1-\frac{1}{2}z^{-1}} + \frac{1}{1-2z^{-1}} ; |z| > 2$



o $\Rightarrow$ zeroes
x $\Rightarrow$ poles



$$|z| > \frac{1}{2}$$

$$\frac{1}{1 - \frac{1}{2}z^{-1}} \Rightarrow \frac{1}{1 - \frac{1}{2z}} = \frac{z}{z - \frac{1}{2}}$$

$N(z)$ zeroes
 $D(z)$ poles

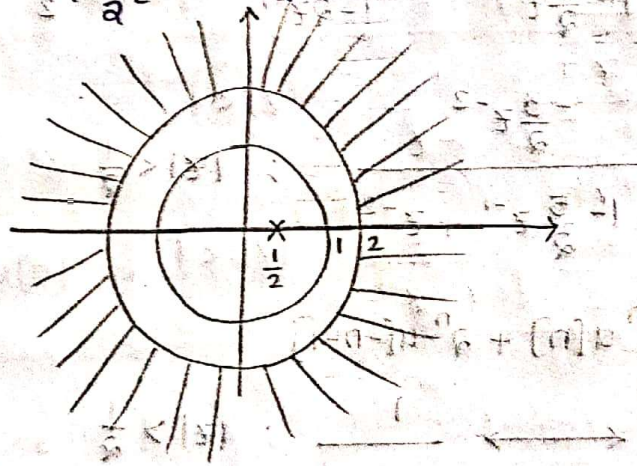
$$|z| > 2$$

$$\frac{1}{1 - 2z^{-1}} = \frac{z}{z - 2}$$

$N(z)$ zeroes
 $D(z)$ poles

$N(z) = 0 \Rightarrow z = 0$
 $D(z) = 0 \Rightarrow z = 2$

$$X(z) = \frac{1}{1 - \frac{1}{2}z^{-1}} + \frac{1}{1 - 2z^{-1}} \quad ; \quad |z| > 2$$



(Pb) Determine the z.T of $x[n] = -(\frac{1}{2})^n u[-n-1] + 2^n u[-n-1]$
Roc, location of poles & zeroes.

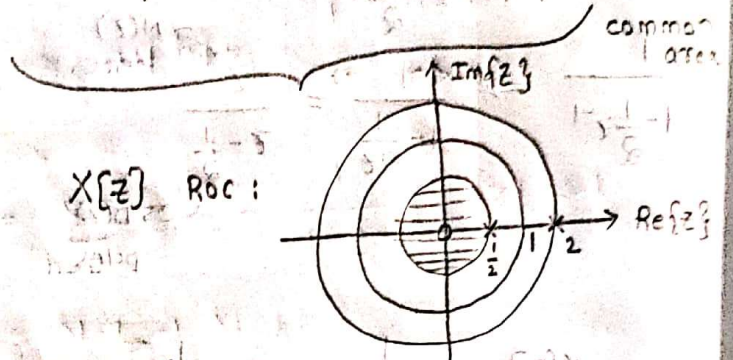
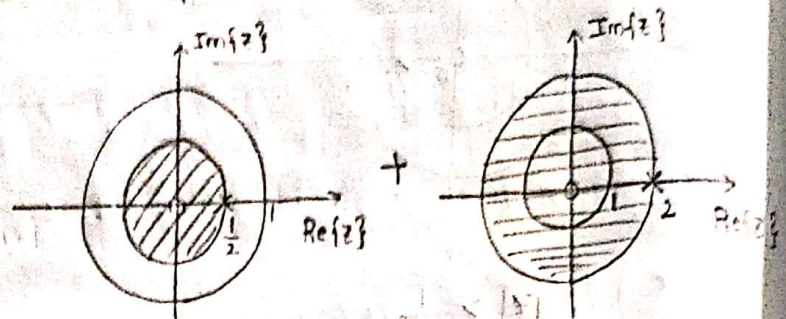
Ans:

$$-a^n u[-n-1] \xleftrightarrow{z.T} \frac{1}{1 - az^{-1}} \quad |z| < |a|$$

$$-(\frac{1}{2})^n u[-n-1] \xleftrightarrow{z.T} \frac{1}{1 - \frac{1}{2}z^{-1}} \quad |z| < |\frac{1}{2}|$$

$$-(2)^n u[-n-1] \xleftrightarrow{z.T} \frac{1}{1 - 2z^{-1}} \quad |z| < 2$$

$$Z\{x[n]\} = X[z] = \frac{1}{1 - \frac{1}{2}z^{-1}} - \frac{1}{1 - 2z^{-1}} ; |z| < \frac{1}{2}$$



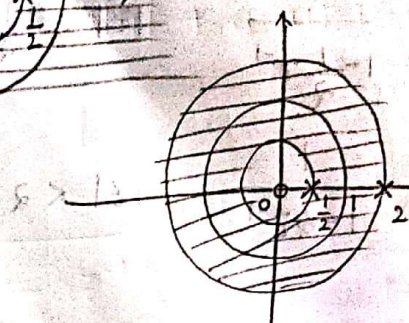
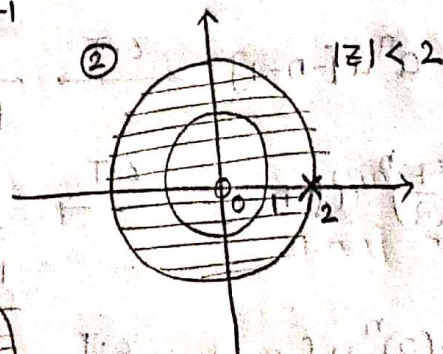
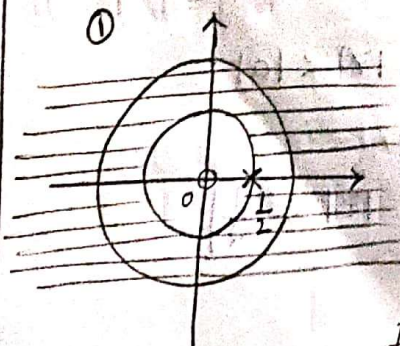
$$X[z] = \frac{1}{1 - \frac{1}{2}z^{-1}} - \frac{1}{1 - 2z^{-1}} ; |z| < \frac{1}{2}$$

$$X[z] = \frac{-\frac{3}{2}z^{-2}}{1 - \frac{5}{2}z^{-1} + z^{-2}} ; |z| < \frac{1}{2}$$

(Pb) $x[n] = (\frac{1}{2})^n u[n] + 2^n u[-n-1]$

Ans: ① $(\frac{1}{2})^n u[n] \longleftrightarrow \frac{1}{1 - \frac{1}{2}z^{-1}} ; |z| > \frac{1}{2}$

② $-2^n u[-n-1] \longleftrightarrow \frac{1}{1 - 2z^{-1}} ; |z| < 2$



(Pb)

Find the z -transform of unit impulse function $x[n] = \delta[n]$.

Ans:

By definition,

$$\begin{aligned}
 z\{x[n]\} = X[z] &= \sum_{n=-\infty}^{\infty} x[n] z^{-n} \\
 &= \sum_{n=-\infty}^{\infty} \delta[n] \cdot z^{-n} = \left. z^{-n} \right|_{n=0} = z^0 = 1
 \end{aligned}$$

↪ exists at origin "n=0"

$$X[z] = 1$$

$$\delta[n] \xleftrightarrow{z.T} 1$$

Roc : Entire z -plane including $z=0$ & $z=\infty$

(Pb)

$$x[n] = u[n]$$

$$\begin{aligned}
 X[z] &= \sum_{n=-\infty}^{\infty} u[n] z^{-n} = \sum_{n=0}^{\infty} z^{-n} = \sum_{n=0}^{\infty} (z^{-1})^n \\
 &= \frac{1}{1 - z^{-1}} \quad |z| > 1
 \end{aligned}$$

$$u[n] \xleftrightarrow{z.T} \frac{1}{1 - z^{-1}} \quad |z| > 1$$

(Pb)

$$x[n] = -u[-n-1]$$

$$X[z] = \sum_{n=-\infty}^{\infty} -u[-n-1] \cdot z^{-n}$$

$$u[n] = \begin{cases} 1 & n \geq 0 \\ 0 & n < 0 \end{cases} \quad u[-n-1] = \begin{cases} 1 & (-n-1) \geq 0 \Rightarrow n \leq -1 \\ 0 & (-n-1) < 0 \Rightarrow n > -1 \end{cases}$$

$$\begin{aligned}
 X[z] &= \sum_{n=-\infty}^{\infty} -u[-n-1] z^{-n} = - \sum_{n=-\infty}^{\infty} z^{-n} \\
 &= - \sum_{n=-\infty}^{\infty} z^{+n} = - \frac{z}{z-1} = - \frac{1}{1-z}
 \end{aligned}$$

$$-u[-n-1] \xleftrightarrow{z \cdot T} \frac{1}{1-z^{-1}} ; |z| < 1$$

(Pb) $x[n] = u[-n]$

$$X[z] = \sum_{n=-\infty}^{\infty} u[-n] \cdot z^{-n}$$

$$u[n] = \begin{cases} 1 & n \geq 0 \\ 0 & n < 0 \end{cases} \quad u[-n] = \begin{cases} 1 & n \leq 0 \\ 0 & n > 0 \end{cases}$$

$$X[z] = \sum_{n=-\infty}^0 z^{-n} = \sum_{n=0}^{\infty} z^n = \frac{1}{1-z} ; |z| < 1$$

$$X[z] = \frac{1}{z[1-z^{-1}]} = \frac{1}{1-z} = \frac{z^{-1}}{1-z^{-1}}$$

$$\begin{aligned} u[n] &\xleftrightarrow{z \cdot T} \frac{1}{1-z} ; |z| < 1 \\ u[-n] &\xleftrightarrow{z \cdot T} \frac{1}{1-z^{-1}} ; |z| < 1 \end{aligned}$$

(Pb) $x[n] = a^{|n|}$

① $0 < a < 1$ ② $1 < a < \infty$

Ans: ① $x[n] = a^{|n|} = \begin{cases} a^n & n \geq 0 \\ a^{-n} & n < 0 \end{cases}$

$$x[n] = \underbrace{a^n u[n]}_{\downarrow} + \underbrace{a^{-n} u[-n-1]}_{\downarrow}$$

$u[n] \rightarrow$ Right side
 $u[-n-1] \rightarrow$ Left side

$$\frac{1}{1-az^{-1}} ; |z| > a \qquad \frac{1}{1-a^{-1}z^{-1}} ; |z| < a$$

$$a^n u[n] \longleftrightarrow \frac{1}{1-az^{-1}} \quad |z| > a$$

$$-a^n u[-n-1] \longleftrightarrow \frac{1}{1-a^{-1}z^{-1}} \quad |z| < \frac{1}{a}$$

$$-a^{-n} u[-n-1] \longleftrightarrow \frac{1}{1-a^{-1}z^{-1}} \quad |z| < \frac{1}{a}$$

$$X(z) = \frac{1}{1-az^{-1}} - \frac{1}{1-a^{-1}z^{-1}} \quad ; a < |z| < \frac{1}{a}$$

$$X(z) = \left(\frac{a^2-1}{a} \right) \frac{z}{(z-a)(z-\frac{1}{a})} \quad ; a < |z| < \frac{1}{a}$$

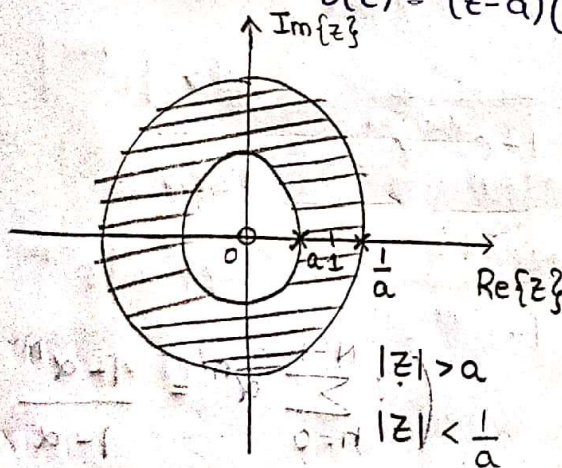
$$K = \frac{a^2-1}{a}$$

$$N(z) = z$$

$$D(z) = (z-a)(z-\frac{1}{a})$$

$$\Rightarrow \text{zeros at } z=0$$

$$\Rightarrow \text{Poles at } z=a, \frac{1}{a}$$



$$a = 0.5$$

$$\left(\frac{1}{a} \right) = \frac{K}{0.12} = 2$$

$$\text{ROC} \Rightarrow a < |z| < \frac{1}{a}$$

$$(b) \quad x[n] = a^{|n|} \quad ; 1 < a < \infty$$

$$= \begin{cases} a^n & n \geq 0 \\ a^{-n} & n < 0 \end{cases}$$

$$x[n] = a^n u[n] - [-a^n u[-n-1]]$$

$$a^n u[n] \longleftrightarrow \frac{1}{1-az^{-1}} \quad |z| > a$$

$$-a^{-n} u[-n-1] \longleftrightarrow \frac{1}{1-a^{-1}z^{-1}} \quad |z| < \frac{1}{a}$$

$$a > 1 \Rightarrow a = 2$$

$$\frac{1}{a} = \frac{1}{2} = 0.5$$

Common Roc doesnot exists,

$\therefore$ Z-Transform won't exist for $|x[n]| = a^n$; $1 < a < \infty$

a^n for $0 < a < 1 \Rightarrow$ Z.T exists

a^n for $1 < a < \infty \Rightarrow$ Z.T doesn't exists.

(Pb) $x[n] = a^n$ $0 \leq n \leq N-1$; $X[z] = ?$

$$X[z] = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

$$= \sum_{n=0}^{N-1} a^n z^{-n}$$

$$= \sum_{n=0}^{N-1} (az^{-1})^n$$

$$= \frac{1 - (az^{-1})^N}{1 - az^{-1}}$$

$$\left(\because \sum_{n=0}^{N-1} \alpha^n = \frac{1 - \alpha^N}{1 - \alpha} \right)$$

$$X[z] = \frac{1}{z^{N-1}} \cdot \frac{z^N - a^N}{z - a}$$

(Pb) $x[n] = \begin{cases} 1 & 0 \leq n \leq N-1 \\ 0 & \text{otherwise} \end{cases}$

(Pb) $x[n] = \sin(\omega_0 n) u[n]$

Ans:- $x[n] = \left[\frac{e^{j\omega_0 n} - e^{-j\omega_0 n}}{2j} \right] u[n]$

$$z\{x[n]\} = \frac{1}{2j} z \left\{ e^{j\omega_0 n} u[n] \right\} - \frac{1}{2j} z \left\{ e^{-j\omega_0 n} u[n] \right\}$$

We know that, $a^n u[n] \longleftrightarrow \frac{1}{1-az^{-1}} \quad |z| > |a|$

① $(e^{j\omega_0})^n u[n] \longleftrightarrow \frac{1}{1 - e^{j\omega_0} z^{-1}} \quad |z| > |e^{j\omega_0}|$
 $|z| > 1$

② $(e^{-j\omega_0})^n u[n] \longleftrightarrow \frac{1}{1 - e^{-j\omega_0} z^{-1}} \quad |z| > |e^{-j\omega_0}|$
 $|z| > 1$

$$X[z] = \frac{1}{2j} \left[\frac{1}{1 - e^{j\omega_0} z^{-1}} - \frac{1}{1 - e^{-j\omega_0} z^{-1}} \right] \quad |z| > 1$$

$$X[z] = \frac{1}{2j} \left[\frac{1 - e^{-j\omega_0} z^{-1} - 1 + e^{j\omega_0} z^{-1}}{(1 - e^{j\omega_0} z^{-1})(1 - e^{-j\omega_0} z^{-1})} \right] \quad |z| > 1$$

$$X[z] = \frac{z^{-1} \left[\frac{e^{j\omega_0} - e^{-j\omega_0}}{2j} \right]}{1 - az^{-1} \left[\frac{e^{j\omega_0} + e^{-j\omega_0}}{2} \right] + z^{-2}} \quad |z| > 1$$

$$X(z) = \frac{z^{-1} \sin(\omega_0)}{1 - 2z^{-1} \cos(\omega_0) + z^{-2}} \quad |z| > 1$$

$$\sin(\omega_0 n) u[n] \xleftrightarrow{z.T} \frac{z^{-1} \sin(\omega_0)}{1 - 2z^{-1} \cos(\omega_0) + z^{-2}} \quad |z| > 1$$

(Pb) $x[n] = \cos(\omega_0 n) u[n]$

$$x[n] = \left[\frac{e^{j\omega_0 n} + e^{-j\omega_0 n}}{2} \right] u[n]$$

$$Z\{x[n]\} = \frac{1}{2} Z\{e^{j\omega_0 n} u[n]\} + \frac{1}{2} Z\{e^{-j\omega_0 n} u[n]\}$$

We know that, $a^n u[n] \longleftrightarrow \frac{1}{1 - az^{-1}} \quad |z| > |a|$

$$(e^{j\omega_0})^n u[n] \longleftrightarrow \frac{1}{1 - e^{j\omega_0} z^{-1}} \quad |z| > |e^{j\omega_0}|$$

$$(e^{-j\omega_0})^n u[n] \longleftrightarrow \frac{1}{1 - e^{-j\omega_0} z^{-1}} \quad |z| > |e^{-j\omega_0}|$$

$$X(z) = \frac{1}{2} \left[\frac{1}{1 - e^{j\omega_0} z^{-1}} + \frac{1}{1 - e^{-j\omega_0} z^{-1}} \right] \quad |z| > 1$$

$$X(z) = \frac{1}{2} \left[\frac{1 - e^{-j\omega_0} z^{-1} + 1 - e^{j\omega_0} z^{-1}}{(1 - e^{j\omega_0} z^{-1})(1 - e^{-j\omega_0} z^{-1})} \right]$$

$$= \frac{1}{2} \left[\frac{2 - (e^{j\omega_0} + e^{-j\omega_0}) z^{-1}}{(1 - e^{j\omega_0} z^{-1})(1 - e^{-j\omega_0} z^{-1})} \right]$$

$$= \frac{1 - z^{-1} \cos(\omega_0)}{1 - 2z^{-1} \cos \omega_0 + z^{-2}} \quad ; |z| > 1$$

$$\cos(\omega_0 n) u[n] \xleftrightarrow{z.T} \frac{1 - z^{-1} \cos(\omega_0)}{1 - 2z^{-1} \cos(\omega_0) + z^{-2}} ; |z| > 1$$

(pb) Determine the z -Transform & Roc of the following finite duration signals.

(a) $x[n] = \{1, 2, 6, -2, 0, 3\}$

$\uparrow$
 $n=0 \quad 1 \quad 2 \quad 3 \quad 4 \quad 5$

Ans- By definition,

$$X[z] = \sum_{n=-\infty}^{\infty} x[n] z^{-n} = \sum_{n=0}^5 x[n] z^{-n}$$

$$X[z] = x[0]z^0 + x[1]z^{-1} + x[2]z^{-2} + x[3]z^{-3} + x[4]z^{-4} + x[5]z^{-5}$$

$$X[z] = 1 + 2z^{-1} + 6z^{-2} - 2z^{-3} + 0 + 3z^{-5}$$

Roc: Entire z -plane except $z=0$

$\therefore X[z]$ becomes unbounded for $z=0$.

(b) $x[n] = \{1, 2, 6, -2, 0, 3\}$

$\uparrow$
 $n=0 \quad 1 \quad 2 \quad 3 \quad 4 \quad 5$

$$X[z] = 1 \cdot z^2 + 2z^1 + 6z^0 - 2z^{-1} + 0 + 3z^{-3}$$

Roc: Entire z -plane except $z=0$ & $z=\infty$

$\therefore X[z]$ becomes unbounded for both $z=0$ & $z=\infty$

(c) $x[n] = \{0, 0, 1, 2, 6, -2, 3\}$

$\uparrow$
 $n=0 \quad 1 \quad 2 \quad 3 \quad 4 \quad 5 \quad 6$

$$X[z] = z^{-2} + 2z^{-3} + 6z^{-4} - 2z^{-5} + 3z^{-6}$$

Roc: Entire z -plane, except $z=0$

$\therefore X[z] = \infty$ for $z=0$.

$$(d) x[n] = \{1, 2, 6, -2, 0, 3\}$$

$$X(z) = z^5 + 2z^4 + 6z^3 - 2z^2 + 3$$

Roc is entire z -plane except $z = \infty$

$\therefore X(z)$ becomes unbounded for $z = \infty$.

Properties of z -Transform:

1. Linearity:

If $x_1[n] \longleftrightarrow X_1(z)$ with $\text{Roc} = R_1$

and $x_2[n] \longleftrightarrow X_2(z)$ with $\text{Roc} = R_2$

then $ax_1[n] + bx_2[n] \longleftrightarrow aX_1(z) + bX_2(z)$
with Roc containing $R_1 \cap R_2$

2. Time shifting:

If $x[n] \longleftrightarrow X(z)$ with $\text{Roc} = R$

then $x[n - n_0] \longleftrightarrow z^{-n_0} X(z)$

with $\text{Roc} = R$, except for the possible addition and deletion of $z = 0$ or $z = \infty$

3. Scaling in z -domain:

If $x[n] \longleftrightarrow X(z)$ with $\text{Roc} = R$

then $z_0^n x[n] \longleftrightarrow X\left(\frac{z}{z_0}\right)$ with $\text{Roc} = |z_0| R$

' z_0 ' scaling factor

4. Time Reversal:

If $x[n] \longleftrightarrow X(z)$ with $\text{Roc} = R$

then $x[-n] \longleftrightarrow X\left(\frac{1}{z}\right) = X(z^{-1})$ with $\text{Roc} = \frac{1}{R}$

5. Differentiation in z-domain:

If $x[n] \longleftrightarrow X[z]$ with $\text{Roc} = R$

then $nx[n] \longleftrightarrow -z \frac{dX[z]}{dz}$ with $\text{Roc} = R$

6. Time Expansion:

Let 'm' be a positive integer

$x_{<m>}[n] = \begin{cases} x[\frac{n}{m}] & \text{if } n \text{ is a multiple of } m \\ 0 & \text{if } n \text{ is not a multiple of } m \end{cases}$

$x_{<m>}[N]$ is obtained from $x[n]$ by placing $(m-1)$ zeroes between successive values of the original signal

$x_{<m>}[N] \Rightarrow$ Slowed version of $x[n]$

If $x[n] \longleftrightarrow X[z]$ with $\text{Roc} = R$

then $x_{<m>}[n] \longleftrightarrow X[z^m]$ with $\text{Roc} = R^m$

7. Convolution:

If $x_1[n] \longleftrightarrow X_1[z]$ with $\text{Roc} = R_1$

$x_2[n] \longleftrightarrow X_2[z]$ with $\text{Roc} = R_2$

then $x_1[n] * x_2[n] \longleftrightarrow X_1[z] \cdot X_2[z]$

with Roc containing $R_1 \cap R_2$

Prop-6: Example:

$x[n] = \{1, 2, 3, 4\}$

$x_{<2>}[n] = x[\frac{n}{2}]$

$(m-1) = (2-1)$ zeroes should be placed b/w two successive samples

$x_{<2>}[n] = \{1, 0, 2, 0, 3, 0, 4\}$ ($\because$ Padding zeroes)

8. Accumulation:

If $x[n] \longleftrightarrow X[z]$ with $\text{Roc} = R$

then $\sum_{k=-\infty}^n x[k] \longleftrightarrow \frac{1}{1-z^{-1}} X[z]$
with Roc atleast $R \cap \{|z| > 1\}$

9. First difference:

If $x[n] \longleftrightarrow X[z]$ with $\text{Roc} = R$

then $x[n] - x[n-1] \longleftrightarrow (1-z^{-1}) X[z]$
with Roc atleast $R \cap \{|z| > 0\}$

10. Conjugation and Conjugate Symmetry:

If $x[n] \longleftrightarrow X[z]$ with $\text{Roc} = R$

then $x^*[n] \longleftrightarrow X^*[z^*]$ with $\text{Roc} = R$.

Ex:1 Determine the Z.T & Roc of the signal

$$x[n] = a^n u[n] - a^n u[n-1]$$

Ans:- By definition,

$$\begin{aligned} X[z] &= \sum_{n=-\infty}^{\infty} x[n] z^{-n} \\ &= \sum_{n=-\infty}^{\infty} a^n u[n] z^{-n} - \sum_{n=-\infty}^{\infty} a^n u[n-1] z^{-n} \\ &= \sum_{n=0}^{\infty} (az^{-1})^n - \sum_{n=0}^{\infty} (az^{-1})^n \\ &= \frac{1}{1-az^{-1}} - \frac{az^{-1}}{1-az^{-1}} \\ &= \frac{1-az^{-1}}{1-az^{-1}} = 1 \Rightarrow X[z] = 1 \end{aligned}$$

Roc is entire z -plane

Note: The sequence $a^n u[n]$ & $a^n u[n-1]$ both have Roc defined by $|z| > |a|$

But the signal $x[n] = a^n u[n] - a^n u[n-1] = \delta[n]$ is finite duration signal. $\{\delta[n]\}$



& its z-T is $X(z) = 1$ which has Roc i.e. the entire z-plane

$$\begin{aligned} a^n [u[n] - u[n-1]] &= a^n \delta[n] \\ &= a^n \delta[n] \Big|_{n=0} \\ &= a^0 \delta[n] \\ &= \delta[n] \end{aligned}$$

Q2 Determine the z-T & Roc of the following signals.

(a) $x[n] = \delta[n-k]$

We know that, $\delta[n] \longleftrightarrow 1$ Roc: Entire z-plane

$$\delta[n-k] \longleftrightarrow 1 \cdot z^{-k} \quad \text{Roc: Entire z-plane except } z=0$$

(b) $x[n] = \delta[n+k]$

$$\delta[n] \longleftrightarrow 1$$

Roc: Entire z-plane

$$\delta[n+k] \longleftrightarrow 1 \cdot z^k \quad \text{Roc: Entire z-plane except } z=\infty$$

(c) $x[n] = 2\delta[n+2] + 3\delta[n] - 5\delta[n-1] + 3\delta[n-2]$

$$X(z) = 2z^2 + 3 - 5z^{-1} + 3z^{-2}$$

Roc: Entire z-plane except $z=0$ & $z=\infty$

(d) $x[n] = u[n] - u[n-10]$

$$X(z) = \frac{1}{1-z^{-1}} - \frac{z^{-10}}{1-z^{-1}}$$

$$= \frac{1-z^{-10}}{1-z^{-1}} = \frac{z}{z^{10}} \left[\frac{z^{10}-1}{z-1} \right] = \frac{1}{z} \left[\frac{z^{10}-1}{z-1} \right]$$

Roc: Entire z-plane except $z=0$